

7.6 Extrinsic Information Transfer (EXIT)

7.6.1. Introduction to EXIT

- In each iteration of the turbo code decoding, the extrinsic information is updated while the channel LLRs remain fixed. It is obvious that, for the decoder to converge, the extrinsic information should be providing improved information about the transmitted bits at each iteration.
- The mutual information $I(X;E)$ between the extrinsic information E and the transmitted symbols x can be used to measure the usefulness of the extrinsic information A .
- Also it is known that the bit error rate curves for turbo codes have a distinct waterfall region : once E_b/N_0 crosses a threshold, the BER decays sharply.

- **Extrinsic information transfer (EXIT) chart provide a computationally efficient tool for visualizing the progress of iteration of iterative decoding , showing what happen before and after the waterfall region.**

7.6.2 Mutual Information

- **The mutual information between two random variables, X and Y , is expressed by**

$$I(X;Y) = H(X) - H(X | Y) \quad (7.44)$$

The mutual information $I(X;Y)$ gives the amount of uncertainty in X that is removed by knowing Y . It can be calculated as

$$I(X;Y) = \sum_x \sum_y p(x,y) \log_2 p(x,y) / p(x)p(y) \quad (7.45)$$

When $p(x,y) = p(x) p(y)$, $I(X;Y) = 0$

- The binary-input additive white Gaussian noise (BI-AWGN) channel can be described by the equation

$$y_i = \mu x_i + n_i$$

where x_i is the transmitted symbol, y_i is the received symbol and n_i is the additive Gaussian noise with zero-mean and variance $\sigma^2 = 2\mu$.

The probability density function (pdf) for n_i is

$$\begin{aligned} p(n_i) &= \{1/\sqrt{(2\pi\sigma^2)}\} \exp(-n_i^2/2\sigma^2) \\ &= \{1/\sqrt{(2\pi\sigma^2)}\} \exp(-n_i^2/2\sigma^2) \end{aligned} \quad (7.46)$$

The mutual information $I(X;Y)$ is then

$$\begin{aligned} I(X;Y) &= \sum_x \int p(x,y) \log_2 \{p(x,y)/p(x)p(y)\} dy \\ &= \sum_x \int p(y|x) p(x) \log_2 \{p(y|x)/p(y)\} dy \end{aligned} \quad (7.47)$$

the conditional pdf $p(y|x)$ is given by

$$p(y|x) = 1/\sqrt{(2\pi\sigma^2)} \exp\{-[y - (\sigma^2 x / 2)]^2 / 2\sigma^2\} \quad (7.48)$$

Then we obtain

$$I(X;Y) = 1 - \left\{ \frac{1}{\sigma\sqrt{2\pi}} \int \exp\left[-\frac{(y-\sigma^2/2)^2}{2\sigma^2}\right] \log_2(1+e^{-y}) dy \right\} \quad (7.49)$$

7.6.3. EXIT Chart

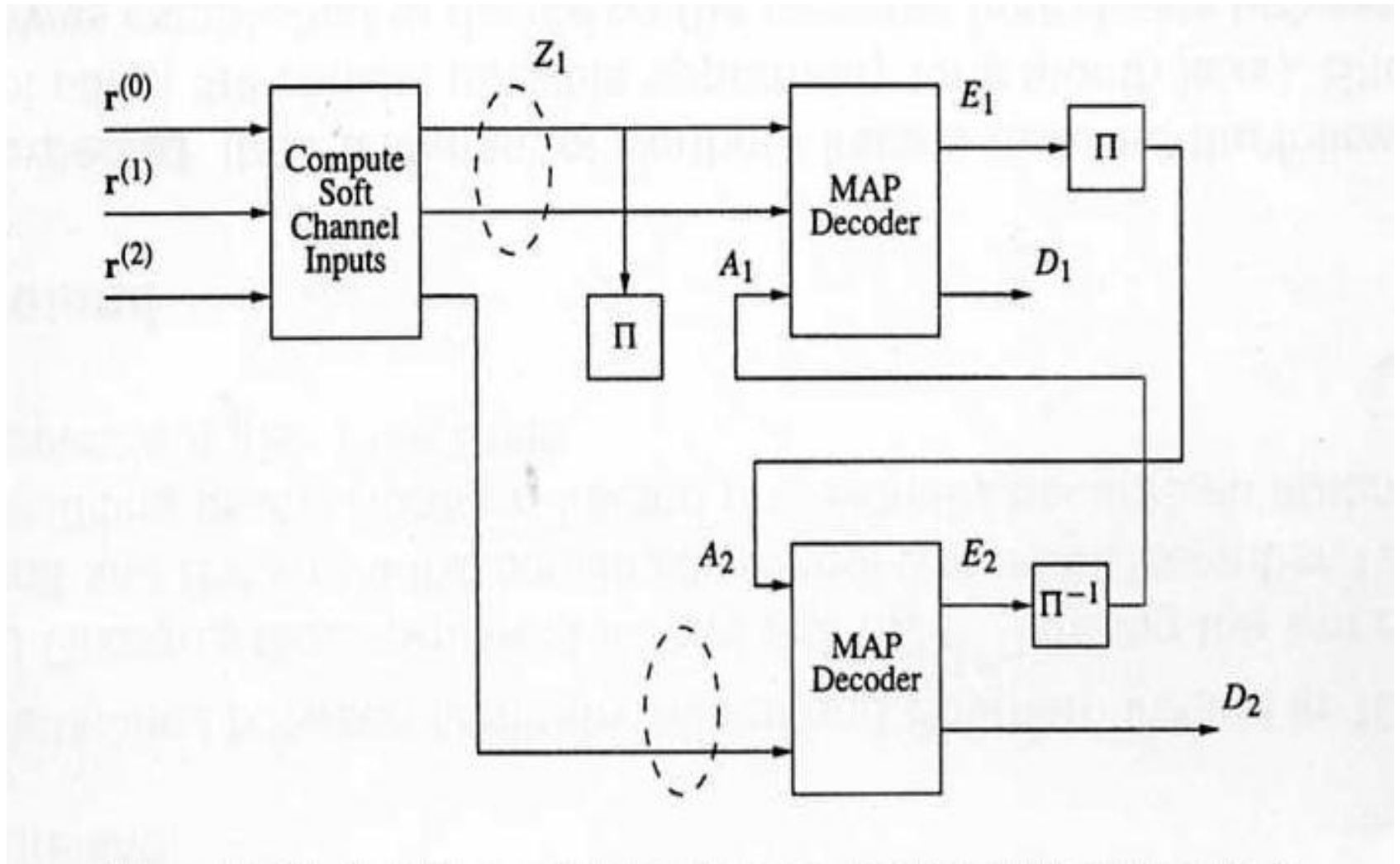
- Fig.7.xx shows an iterative decoder for parallel-concatenated codes. The variables used for EXIT chart analysis are also shown in the figure. Two MAP decoders are used in the iterative decoder.

- The a priori information is labeled as A_i , $i = 1, 2$.

The extrinsic information is E_i , the decoder output information is D_i , and the soft input information is Z_i , with

$$D_i = Z_i + A_i + E_i \quad (7.50)$$

Fig. 7.x1 iterative decoder for parallel-concatenated codes



- The key concept of the EXIT chart is measuring the amount of information that the prior A conveys about the transmitted data X and that the extrinsic information conveys about the data information X . This information is measured using mutual information in the form of $I(X; A_i)$ and $I(X; E_i)$.
- The received discrete-time signal from the AWGN channel is expressed as

$$z = x + n$$

The conditional probability density function $p(z | x)$ is given by

$$p(z | x) = \{1 / \sqrt{(2\pi\sigma_n^2)}\} \exp(-(z-x)^2 / 2\sigma_n^2)$$

- The soft output from the channel is obtained from the LLR of z is

$$Z = \ln \{ p(z | x=1) / p(z | x=-1) \} = 2(x+n) / 2\sigma_n^2$$

- The above equation can also be formulated as

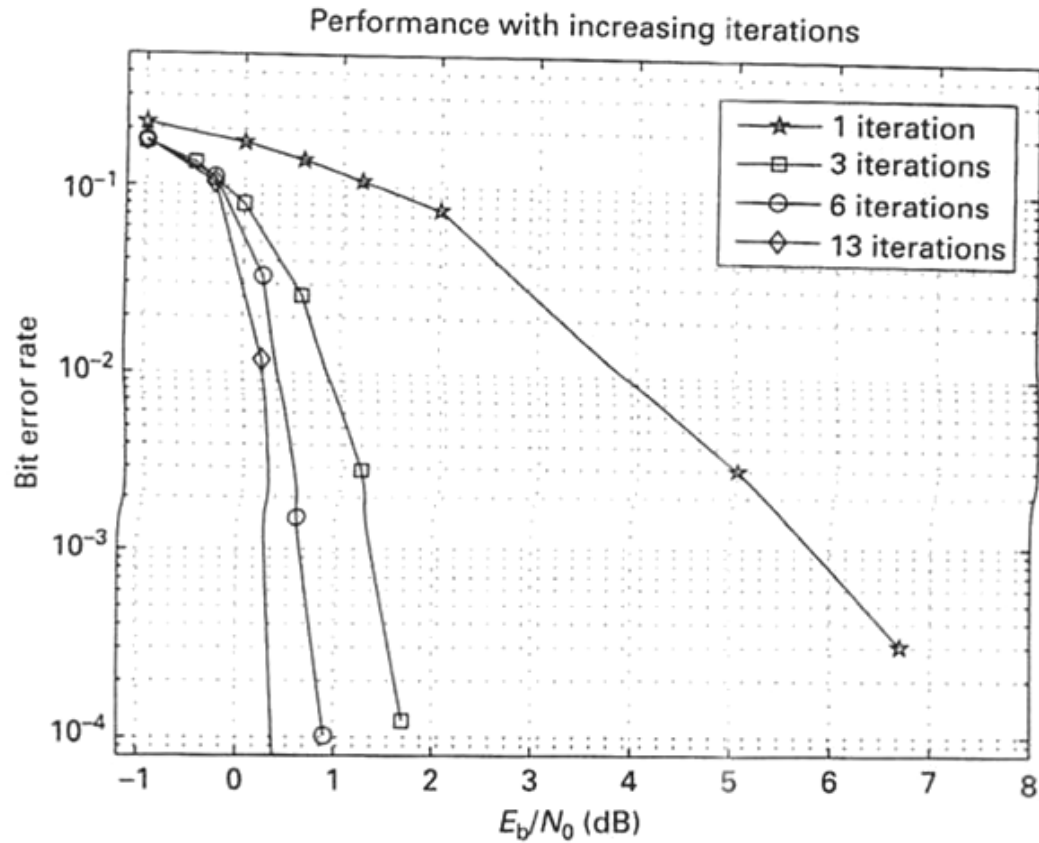
$$Z = \mu_z x + n_z$$

with $\mu_z = 2 / \sigma_n^2$

and n_z being Gaussian distributed with zero-mean and variance $\sigma_z^2 = 4 / \sigma_n^2 = 2 \mu_z$

- The following assumptions are made for the analysis :
 - (1) For sufficiently large interleavers , the a priori values A_i are fairly uncorrelated from their respective channel observations Z_i over many iterations .
 - (2) The probability density function of the extrinsic information E_i –which are the prior inputs for the next MAP decoder – approach Gaussian distribution with increasing iterations.

Bit error rate as a function of number of iterations for a rate 1/3 turbo code



- The mutual information between \mathbf{X} and \mathbf{E} can be computed by

$$I_E = I(\mathbf{X}; \mathbf{E})$$

$$= (1/2) \int_{-\infty}^{\infty} p_E(\mathbf{y} \mid \mathbf{X} = \mathbf{x})$$

$$\log_2 \{ 2 p_E(\mathbf{y} \mid \mathbf{X} = \mathbf{x}) /$$

$$[p_E(\mathbf{y} \mid \mathbf{X} = +1) + p_E(\mathbf{y} \mid \mathbf{X} = -1)] \}$$

$$0 \leq I_E \leq 1$$

- There is some functional relationship between I_A , E_b , and I_E , denoted abstractly as

$$I_E = T(I_A, E_b / N_0)$$

For a fixed E_b / N_0 , we have

$$I_E = T(I_A)$$

The function T denotes the “transfer” of information from the prior information A at the input of a decoder to the extrinsic information E at the output of the decoder.

- **Mutual information $I(X;A)$:**

We can model the a priori probability input to a decoder as

$$A_i = \mu_A x_i + n_A$$

The conditional pdf belonging to the LLR of A is

$$p_A(y \mid \mathbf{X} = \mathbf{x}) \\ = \{1 / \sqrt{(2\pi\sigma_A^2)}\} \exp[-(y - \sigma_A^2/2)^2 / 2\sigma_A^2]$$

Then the mutual information between X and A can be computed by

$$I(X;A) = I_A(\sigma_A) \\ = 1 - \{1 / \sigma \sqrt{(2\pi)} \cdot \\ \int_{-\infty}^{\infty} \exp[-(y - \sigma_A^2/2)^2 / 2\sigma_A^2] \log_2(1+e^{-y}) dy \}$$

For abbreviation we define $J(\sigma) = I_A(\sigma_A = \sigma)$

- Consider the iterative decoder shown in Fig. 7.xx .

The extrinsic information E_1 at the output of the first decoder is permuted and used as the prior information A_2 at the next decoder .

Let $I_{A1}^{[n]}$ denote the mutual information $I(X;A_1)$ at the n th iteration of the first decoder , starting with zero a priori knowledge $I_{A1}^{[0]} = 0$.

Similarly let $I_{E1}^{[n]} = I(X;E_1)$ denote the mutual information $I(X;E_1)$ at the output of the first decode at the n th iteration , $I_{E1}^{[n]} = T_1(I_{A1}^{[n]})$. This is forwarded to to the next decoder to become $I_{A2}^{[n]} = I_{E1}^{[n]}$. This passes through the second decoder to become $I_{E2}^{[n]} = T_2(I_{A2}^{[n]})$, which in turn is passed back to the first decoder as the prior , $I_{A1}^{[n+1]}$.

Fig.7.x2 shows the mutual information I_E at the output of a single decoder as a function of the mutual information I_A at the input of the decoder .

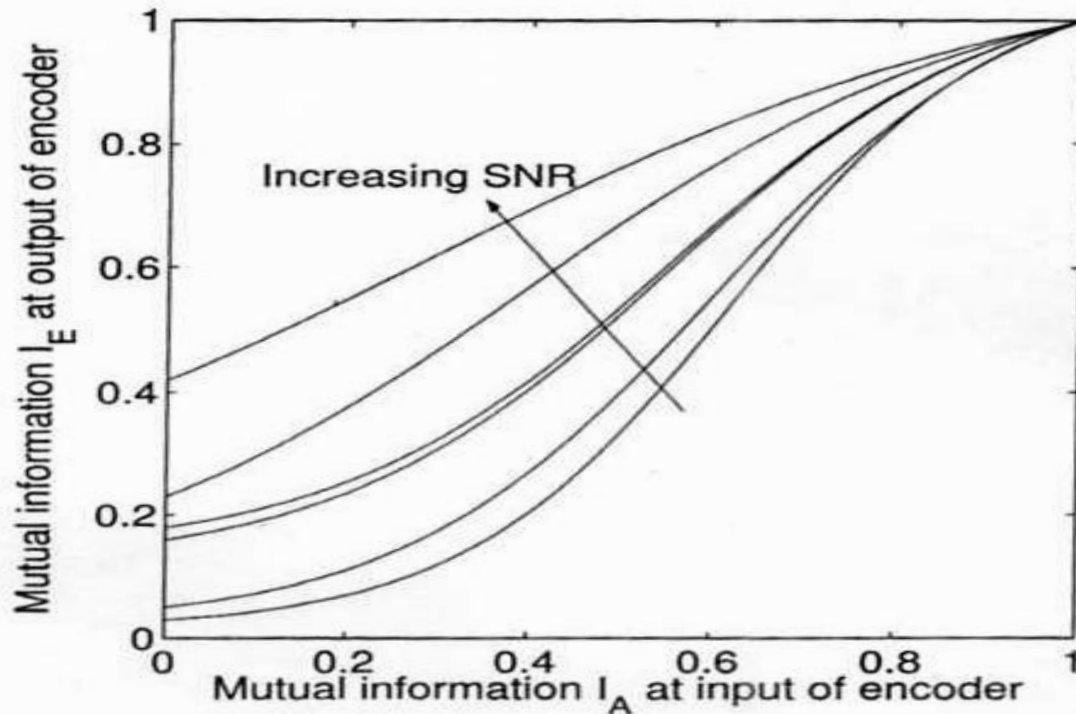
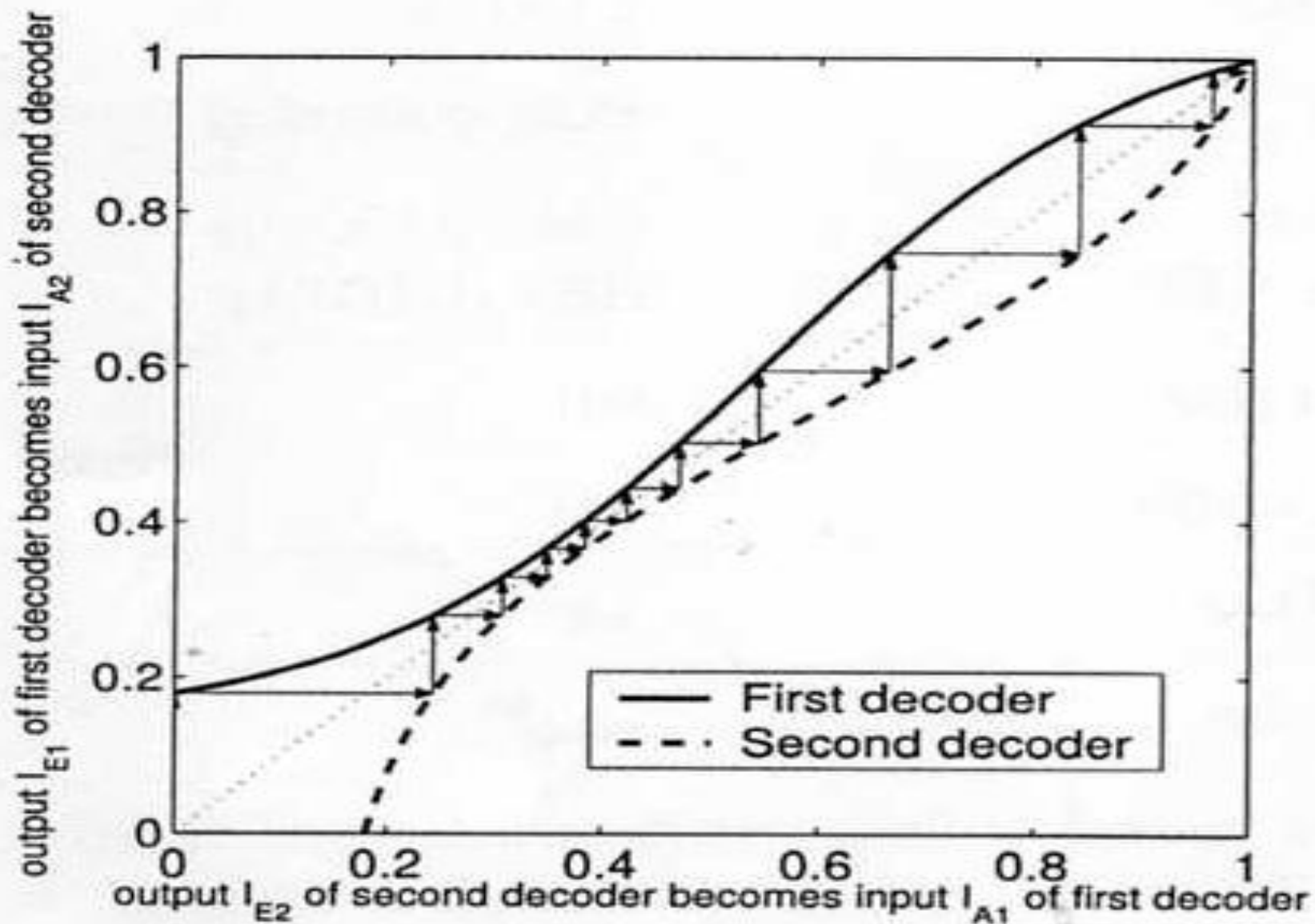


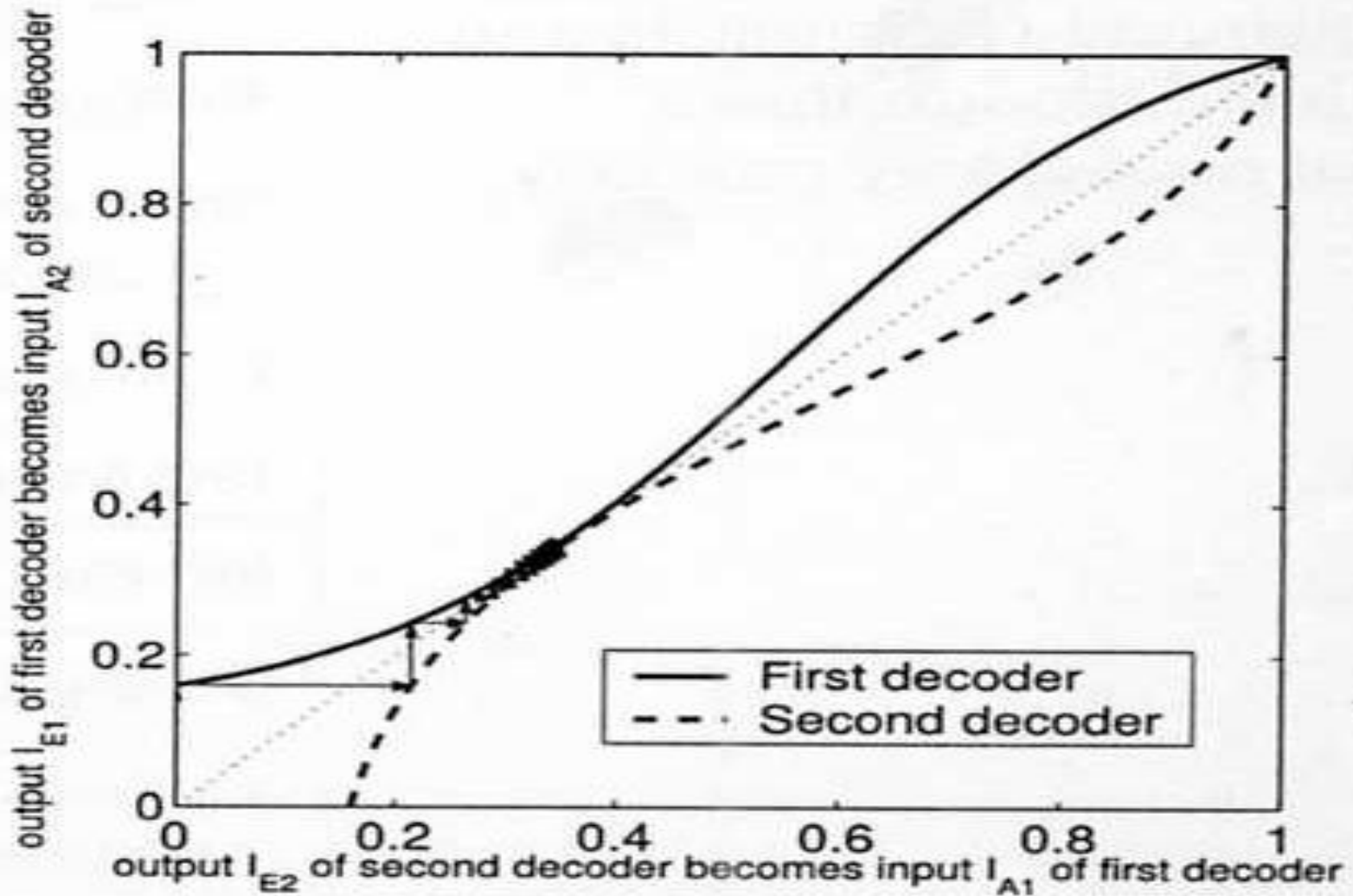
Fig.7.x2a) shows the information decoding in a sequence of decoding steps, following the arrows. Ultimately, a point is reached where $I_A = 1$. If the prior information about a bit is sufficiently close to 1, then we conclude that the prior information is sufficient to accurately decode the bit. There is a “channel” or “gap” between the two curves. The decoding proceeds by walking through this channel.

Fig.7.x2(b) shows that the iterative decoding process can break down for a lower SNR. In this case, the function $T(I_A)$ crosses the $y = x$ line. As a result, the iterations get stuck at the crossover point.

Thus, there is a threshold of SNR for the decoder to be able to decode correctly. The decoder can not correctly decode, no matter how many times the decoder iterates, if the SNR is below the threshold.



(a) Decoding above threshold.



(b) Decoding below threshold. The decoder cannot get past the pinchoff point.

Reference

ten Brink,S., “ Convergence Behaviors of Iteratively Decoded Parallel Concatenated Codes ,” IEEE Trans. Commun. Vol.49, no.10 ,pp.1727-1737, Oct. 2001.