## Homework 3

Due on $11 / \mathbf{1 7} / \mathbf{2 0 0 9}$, Tuesday, before class

## Reading Assignments

Chapter 3 and Chapter 4 of Gallager's note. Sec. 4.8, Sec. 5.8, Sec. 5.9, and Sec. 5.10 of textbook.

Problem Assignments

1. (10 points) Show that the $Q$-function, defined by $Q(x)=P[Z \geq x]=\frac{1}{\sqrt{2 \pi}} \int_{x}^{\infty} e^{-\frac{z^{2}}{2}} d z$ with standard Gaussian random variable $Z$, has the following upper bound for $x \geq 0$.:

$$
Q(x) \leq \frac{1}{2} e^{-\frac{x^{2}}{2}}
$$

2. $(10+10=20$ points) Consider a binary hypothesis testing problem, and denote the hypotheses as $\mathrm{H}_{1}: \alpha=1$ and $\mathrm{H}_{2}: \alpha=-1$. Let $\mathbf{h}$ be an $n \times 1$ real vector and the observation random vector y takes the form

$$
\mathrm{y}=\alpha \cdot \mathbf{h}+\mathrm{z}
$$

where $\mathbf{z} \sim \mathcal{N}\left(\mathbf{0}, \sigma^{2} \mathbf{I}_{n}\right)$ is independent with $\alpha$ and $\mathbf{I}_{n}$ is an $n \times n$ identity matrix.
(a) Find the maximum likelihood decision rule for $\alpha$ and find the probability of error in terms of the $Q$ function.
(b) Now suppose a third hypothesis $\alpha=0$ is added to the situation of part (a). The observation random vector takes the same form, but here $\alpha$ can take on values $-1,0$ or 1 , equally likely. Find the maximum likelihood decision rule in this case, and determine the error probability.
3. $(10 \times 4=40$ points $)$ Let the mathematical model of a communication system with $N$ transmit and $M$ receive antennas be given by

$$
\mathbf{y}_{i}=\sqrt{\frac{E_{s}}{N}} \mathbf{H s}_{i}+\mathbf{w}_{i}
$$

where $\mathbf{y}_{i}=\left[y_{i, 1}, y_{i, 2}, \cdots, y_{i, M}\right]^{T}$ is the $M \times 1$ vector collecting received signals from all receive antennas at the $i$ th time instant, $\mathbf{H}$ is the time-invariant channel matrix with dimension $M \times N, \mathbf{s}_{i}$ is the $N \times 1$ transmitted signal vector (with unit average energy) at the $i$ th time instant, $E_{s}$ is the total energy of the transmitted signal, and $\mathbf{w}_{i} \sim \mathcal{C N}\left(\mathbf{0}, \mathrm{~N}_{0} \mathbf{I}_{M}\right)$ is the complex Gaussian noise vector with $\mathrm{N}_{0}$ being the variance of each noise component and $\mathbf{I}_{M}$ the identity matrix with size $M \times M$. More specifically, the model tells that the $m$ th component $y_{i, m}$ of $\mathbf{y}_{i}$ represents the received signal at the $m$ th antenna of the receiving device, and is equal to

$$
y_{i, m}=\sum_{n=1}^{N} h_{m, n} s_{i, n}+w_{i, m}
$$

where $h_{m, n}$ is the $(m, n)$ th component of $\mathbf{H}$ which means the channel gain between the $n$th transmit antenna and the $m$ th receive antenna, $s_{i, n}$ is the $n$th component of $\mathbf{s}_{i}$ signifying the transmitted signal from the $n$th transmit antenna at the $i$ th time instant, and $w_{i, m}$ is the noise.

We collect $T$ received signal vectors, from $i=1$ to $i=T$, and stack them into

$$
\mathbf{Y}=\sqrt{\frac{E_{s}}{N}} \mathbf{H S}+\mathbf{W}
$$

where $\mathbf{Y}=\left[\mathbf{y}_{1}, \mathbf{y}_{2}, \cdots, \mathbf{y}_{T}\right]$, and likewise for the representations of $\mathbf{S}$ and $\mathbf{W}$. The signal codeword matrix $\mathbf{S}$ belongs to a set $\mathcal{S}=\left\{\mathbf{S}^{(1)}, \mathbf{S}^{(2)}, \cdots, \mathbf{S}^{(K)}\right\}$, called codebook, with size $K$.
(a) Given that the receiver knows the channel matrix $\mathbf{H}$, show that the ML detection rules gives the following decision criterion:

$$
\hat{\mathbf{S}}_{M L}=\arg \min _{\mathbf{S}}\left\|\mathbf{Y}-\sqrt{\frac{E_{s}}{N}} \mathbf{H S}\right\|_{F}^{2}
$$

where the Frobenious norm $\|\mathbf{A}\|_{F}$ of a $p \times q$ matrix $\mathbf{A}$ is defined as the sum of squared absolute values of all components, i.e. $\|\mathbf{A}\|_{F}=\sqrt{\sum_{i=1}^{p} \sum_{j=1}^{q}\left|a_{i, j}\right|^{2}}$, where $a_{i, j}$ is the $(i, j)$ th entry of the matrix $\mathbf{A}$.
(b) With the ML decision rule in part (a), show that the conditional pairwise error probability $P\left(\mathbf{S}^{(i)} \rightarrow \mathbf{S}^{(j)} \mid \mathbf{H}\right)$, which means the probability that the receiver decides the codeword $\mathbf{S}^{(j)}$ while the actual transmitted codeword is $\mathbf{S}^{(i)}$, is given by

$$
P\left(\mathbf{S}^{(i)} \rightarrow \mathbf{S}^{(j)} \mid \mathbf{H}\right)=Q\left(\sqrt{\frac{\rho\left\|\mathbf{H} \mathbf{E}_{i, j}\right\|_{F}^{2}}{2 N}}\right)
$$

where $\rho=\frac{E_{s}}{\mathrm{~N}_{0}}$ is the signal-to-noise ratio (SNR) and $\mathbf{E}_{i, j}=\mathbf{S}^{(i)}-\mathbf{S}^{(j)}$.
(Hint: You can use $\|\mathbf{A}\|_{F}^{2}=\operatorname{Tr}\left(\mathbf{A}^{H} \mathbf{A}\right)$, the trace of $\mathbf{A}^{H} \mathbf{A}$, to simplify the derivation.)
(c) From Problem 3 and part (b) of this problem, the conditional pairwise probability can be bounded by

$$
P\left(\mathbf{S}^{(i)} \rightarrow \mathbf{S}^{(j)} \mid \mathbf{H}\right) \leq e^{-\frac{\rho\left\|\mathbf{H E} \mathbf{E}_{i, j}\right\| \|_{F}^{2}}{4 N}} .
$$

Show that the average PEP can be bounded by

$$
P\left(\mathbf{S}^{(i)} \rightarrow \mathbf{S}^{(j)}\right) \leq\left(\frac{1}{\operatorname{det}\left(\mathbf{I}_{N}+\frac{\rho}{4 N} \mathbf{E}_{i, j} \mathbf{E}_{i, j}^{H}\right)}\right)^{M}
$$

where the average is taken over the channel matrix $\mathbf{H}$ whose components are assumed to be i.i.d. complex Gaussian random variables with zero mean and unit variance, $i . e$. the $(i, j)$ th component $h_{i, j} \sim \mathcal{C N}(0,1)$.
(d) Discuss how the average PEP in part (c) scales with the number of transmit $(N)$ and receive $(M)$ antennas as the SNR goes to infinity.
4. $\left(10+10=20\right.$ points) A gamma random variable $Y_{k}$ with parameter $(k, \lambda)$ can be considered as the sum of $k$ independent exponential random variables $T_{i}$ for $i=1 \cdots k$ with parameter $\lambda$, i.e.

$$
Y_{k}=\sum_{i=1}^{k} T_{i}
$$

The gamma random variable $Y_{k}$ is commonly used to model the total amount of time that one has to wait until the $k$ th Poisson event has occurred. Or, equivalently, $T_{i}$ is used to model the inter-arrival time between the $(i-1)$ th and the $i$ th occurrence of the Poisson event, where the time of the zero-th occurrence is labelled as the origin.
(a) Find the probability distribution function $F_{Y_{k}}(y)=P\left[Y_{k} \leq y\right]$ of $Y_{k}$. (Hint: consider the event that the number of Poisson events occurred in the interval $[0, y]$ is greater than $k$.)
(b) Find the probability density function of $Y_{k}$ by differentiating $F_{Y_{k}}(t)$ in part (a).
5. (10 points) Justify that the sample variance is an unbiased estimator.

## Extra Problems

You do NOT need to turn in your solutions for the following problems. However, I strongly encourage you to work through them.

1. Consider a communication system that is corrupted by unknown interference $Z$ and Gaussian noise $W$. Mathematically, the received signal $Y$ can be modeled as

$$
Y=X+Z+W
$$

where $X$ is the desired signal equally likely to 1 or -1 , and $W \sim \mathcal{N}\left(0, \sigma^{2}\right)$. Assume $X, Z, W$ are mutually independent.
(a) Suppose we model the interference $Z$ as a Gaussian random variable with $Z \sim \mathcal{N}\left(0, A^{2}\right)$. Find the maximum likelihood decision rule for $X$ from the received signal $Y$. What is the probability of error decision? And, how does the error probability behave when the power $A^{2}$ approaches to infinity?
(b) More practically, we now model the interference $Z$ as a binary discrete random variable equally likely to be $A$ or $-A$. Find the maximum likelihood decision rule for $X$ from the received signal $Y$.
(c) In part (b), obtain the ML decision rule when $A$ approaches to infinity. And, calculate the probability of error decision in this large power condition.
(d) Explain intuitively the difference between part (a) and part (c).

